

UF 21 Variable MU R.F. pentode

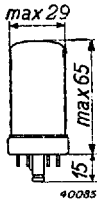


Fig. 1
Dimensions in mm

The UF 21 is a variable-mu R.F. or I.F. amplifier pentode for AC/DC receivers, consuming a heater current of 100 mA. It can also be employed as an R-C coupled A.F. amplifier, in which case it is also possible to obtain very excellent automatic gain control; since in such cases it is very important to know the amount of distortion occurring at a given output voltage and grid bias, details have been included in the Operating Data.

Apart from the heater ratings, the UF 21 is identical with the EF 22, also as far as the sliding screen voltage is concerned.

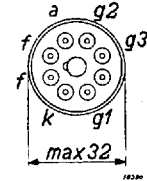
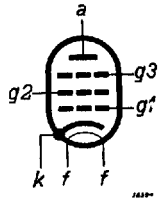


Fig. 2
Arrangement and sequence of contacts.

HEATER RATINGS

Heating: indirect, AC or DC, series supply.

Heater voltage $V_f = 12.6$ V

Heater current $I_f = 0.100$ A

CAPACITANCES

$$C_a = 6.6 \text{ pF}$$

$$C_{g1} = 5.6 \text{ pF}$$

$$C_{ag1} < 0.002 \text{ pF}$$

$$C_{g1f} < 0.006 \text{ pF}$$

OPERATING DATA: valve employed as R.F. or I.F. amplifier

a) With fixed screen voltage.

Anode voltage

$$V_a = 100 \text{ V} \quad 200 \text{ V}$$

Suppressor grid voltage

$$V_{g3} = 0 \text{ V} \quad 0 \text{ V}$$

Screen grid voltage

$$V_{g2} = 100 \text{ V} \quad 100 \text{ V}$$

Cathode resistance

$$R_k = 325 \text{ Ohms} \quad 325 \text{ Ohms}$$

Grid bias

$$V_{g1} = -2.5 \text{ V}^1) -19 \text{ V}^2) \quad -22 \text{ V}^3) \quad -2.5 \text{ V}^1) \quad -19 \text{ V}^2) \quad -22 \text{ V}^3)$$

Anode current

$$I_a = 6 \text{ mA} \quad - \quad - \quad 6 \text{ mA} \quad - \quad -$$

Screen grid current

$$I_{g2} = 1.7 \text{ mA} \quad - \quad - \quad 1.7 \text{ mA} \quad - \quad -$$

Mutual conductance

$$S = 2200 \quad 22 \quad 7 \quad 2200 \quad 22 \quad 7 \mu\text{A/V}$$

Internal resistance

$$R_i = 0.4 \text{ M}\Omega \quad >10 \text{ M}\Omega \quad >10 \text{ M}\Omega \quad 1 \text{ M}\Omega \quad >10 \text{ M}\Omega \quad >10 \text{ M}\Omega$$

Gain factor in respect of screen grid

$$\mu_{g2g1} = 17 \quad - \quad - \quad 17 \quad - \quad -$$

Equivalent noise resistance

$$R_{aeq} = 6200 \text{ Ohms} \quad - \quad - \quad 6200 \text{ Ohms} \quad - \quad -$$

b) With sliding screen voltage.

Anode voltage	$V_a =$	100 V		200 V	
Suppressor grid voltage	$V_{g3} =$	0 V		0 V	
Screen grid resistance	$R_{g2} =$	60,000 Ohms		60,000 Ohms	
Cathode resistance	$R_k =$	325 Ohms		325 Ohms	
Grid bias	$V_{g1} =$	$-1.3 V^1)$ $-19 V^2)$ $-23 V^3)$		$-2.5 V^1)$ $-37 V^2)$ $-46 V^3)$	
Screen grid voltage	$V_{g2} =$	50 V	100 V	100 V	200 V
Anode current	$I_a =$	3.2 mA	—	6 mA	—
Screen grid current	$I_{g2} =$	0.85 mA	—	1.7 mA	—
Mutual conductance	$S =$	2000 20	5	2200 22	$4.5 \mu A/V$
Internal resistance	$R_i =$	$1 M \Omega$	$>10 M \Omega$	$1 M \Omega$	$>10 M \Omega$ $>10 M \Omega$
Equivalent noise resistance	$R_{aeq} =$	4000Ω	—	6200 Ohms	—

1) Valve not controlled.
 2) Mutual conductance controlled to $1/100$.
 3) Extreme limit of control range.

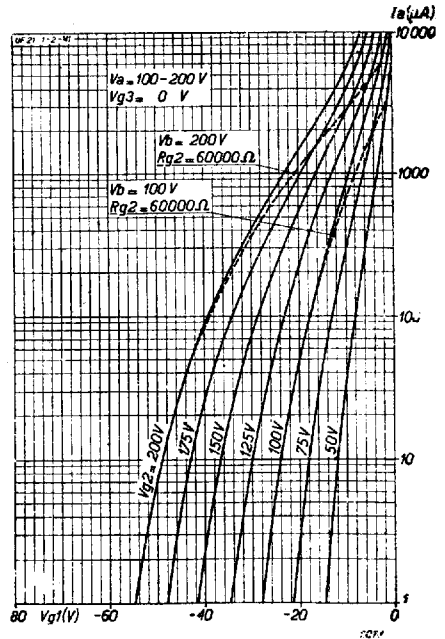


Fig 3
 Anode current as a function of grid bias at $V_a = 100-200 V$ and $V_{g3} = 0 V$, with screen voltage as parameter. The broken lines show the anode current when the valve is controlled, with the screen grid fed through a resistance of 60,000 Ohms from the 200 V or 100 V source.

OPERATING DATA: valve employed as resistance-capacitance coupled A.F. amplifier, with gain control applied to the control grid

Anode coupling res. R_a (M Ohm)	Screen grid res. R_{g_2} (M Ohm)	Anode current I_a (mA)	Screen grid current I_{g_1} (mA)	Cathode res. R_k (Ohms)	Control volts on grid 1. $-V_R$ (V)	Gain	Required alternating grid voltage and total distortion to give an alternating output voltage of:					
							$V_{o_{eff}} = 3\text{ V}$		$V_{o_{eff}} = 5\text{ V}$		$V_{o_{eff}} = 8\text{ V}$	
							$\frac{V_{o_{eff}}}{V_{g_1_{eff}}}$	d_{tot} (%)	$\frac{V_{o_{eff}}}{V_{g_1_{eff}}}$	d_{tot} (%)	$\frac{V_{o_{eff}}}{V_{g_1_{eff}}}$	d_{tot} (%)
$V_b = 200\text{ V}$												
0.2	0.8	0.65	0.17	2500	0	88	0.034	0.75	0.057	1.25	0.091	2.0
0.2	0.8	0.54	0.14	2500	5	35	0.086	1.2	0.140	2.0	0.228	3.2
0.2	0.8	0.46	0.11	2500	10	22	0.136	1.4	0.228	2.3	0.364	3.7
0.2	0.8	0.38	0.08	2500	15	15	0.200	1.7	0.334	2.8	0.534	4.5
0.2	0.8	0.31	0.06	2500	20	11	0.272	1.8	0.455	3.0	0.726	4.8
0.2	0.8	0.25	0.05	2500	25	8	0.375	2.3	0.625	3.8	1.0	5.8
$V_b = 100\text{ V}$												
0.1	0.4	1.2	0.35	1300	0	78	0.038	0.75	0.064	1.25	0.102	2.0
0.1	0.4	0.96	0.28	1300	5	33	0.091	1.2	0.152	2.0	0.242	3.2
0.1	0.4	0.78	0.22	1300	10	20	0.150	1.6	0.250	2.65	0.400	4.25
0.1	0.4	0.62	0.16	1300	15	13	0.230	2.0	0.385	3.3	0.615	5.3
0.1	0.4	0.48	0.12	1300	20	8	0.375	2.2	0.625	3.65	1.000	5.85
0.1	0.4	0.36	0.09	1300	25	6	0.500	3.45	0.832	5.65	1.333	9
$V_b = 100\text{ V}$												
0.2	0.8	0.33	0.08	2500	0	82	0.037	0.85				
0.2	0.8	0.26	0.06	2500	2.5	37	0.081	2.3				
0.2	0.8	0.21	0.055	2500	5	21	0.143	3.4				
0.2	0.8	0.18	0.03	2500	7.5	13	0.230	4.1				
0.2	0.8	0.14	0.025	2500	10	9	0.334	4.3				
0.2	0.8	0.12	0.02	2500	12.5	7	0.430	5.1				
0.1	0.4	0.61	0.15	1300	0	72	0.041	0.85				
0.1	0.4	0.47	0.13	1300	2.5	35	0.086	2.3				
0.1	0.4	0.37	0.10	1300	5	20	0.150	3.45				
0.1	0.4	0.29	0.06	1300	7.5	12	0.250	4.3				
0.1	0.4	0.22	0.05	1300	10	7	0.430	5.25				
0.1	0.4	0.17	0.04	1300	12.5	6	0.500	6.2				

MAXIMUM RATINGS

- V_{a0} ($I_a = 0$) = max. 550 V
- V_a = max. 250 V
- W_a = max. 2 W
- $V_{g_{20}}$ ($I_{g_2} = 0$) = max. 550 V
- V_{g_2} ($I_a < 3\text{ mA}$) = max. 250 V
- V_{g_2} ($I_a = 6\text{ mA}$) = max. 150 V
- W_{g_2} = max. 0.3 W
- I_k = max. 10 mA
- V_{g_1} ($I_{g_1} = +0.3\ \mu\text{A}$) = max. -1.3 V
- $R_{g_{1k}}$ = max. 3 M Ohms
- R_{fk} = max. 20,000 Ohms
- V_{fk} = max. 150 V

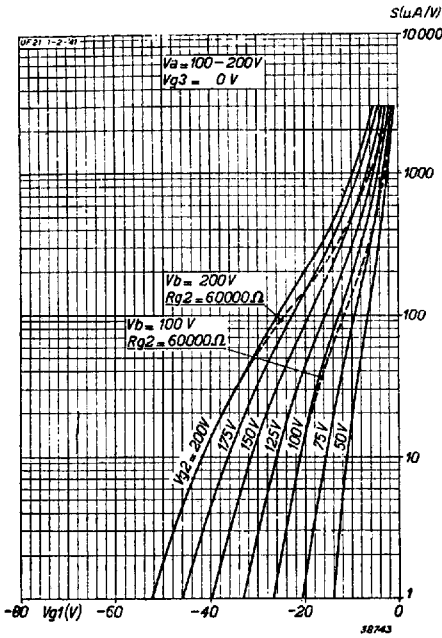


Fig. 4

Mutual conductance as a function of grid bias, with screen grid voltage as parameter. The dotted lines show the conductance with control on the valve and the screen fed through a resistance of 60,000 Ohms from the 100 V or 200 V source.

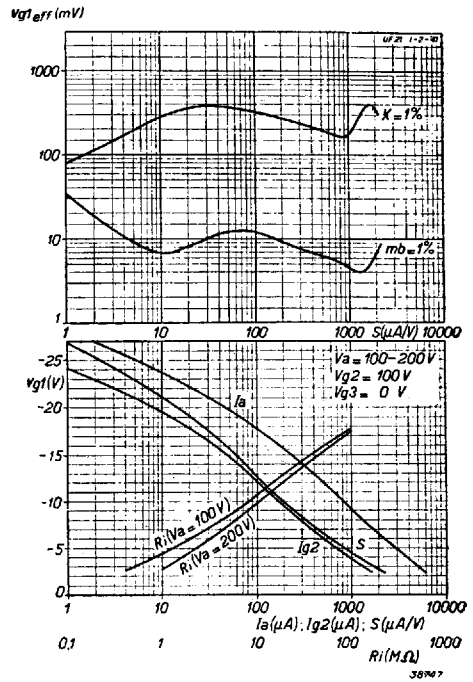


Fig. 5

At $V_a = 100-200V$, $V_{g2} = 100V$ (fixed screen voltage) and $V_{g3} = 0V$. Upper diagram; maximum permissible effective value of alternating grid voltage with 1% cross modulation ($K = 1\%$) and also with 1% modulation hum ($mb = 1\%$), as function of the mutual conductance. Lower diagram; Mutual conductance, anode current, screen grid current and internal resistance as function of grid bias.

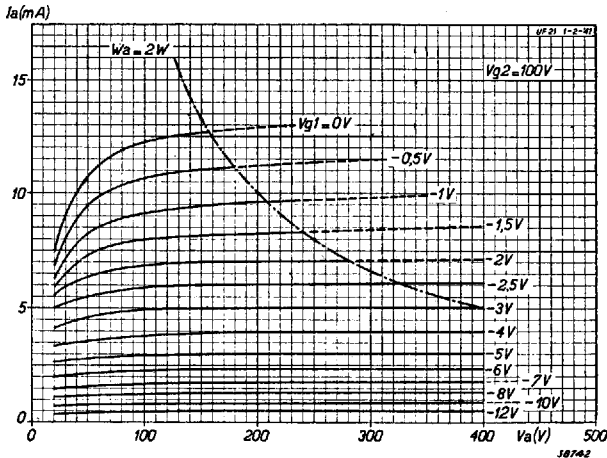


Fig. 6

Anode current as a function of anode voltage at a fixed screen voltage of 100 V, with grid bias as parameter.

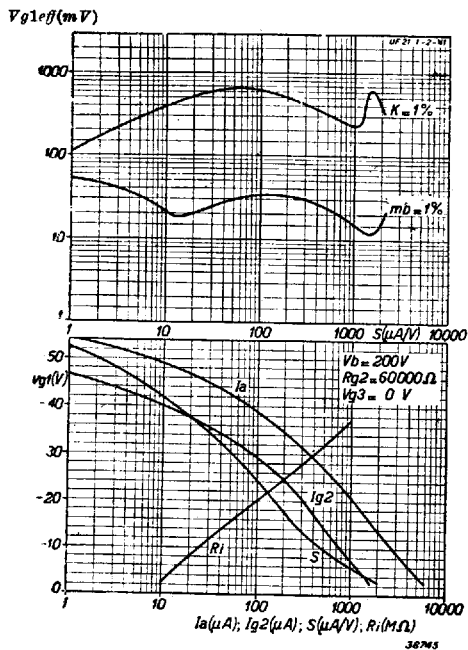


Fig. 7

At $V_b = 200 V$, $R_{g2} = 60,000$ Ohms (screen fed through a resistance) and $V_{g3} = 0 V$. Upper diagram; maximum permissible effective value of alternating grid voltage with 1 % cross modulation ($K = 1\%$) and also 1 % modulation hum ($mb = 1\%$), as function of mutual conductance.

Lower diagram; Mutual conductance, anode current, screen grid current and internal resistance as function of grid bias.

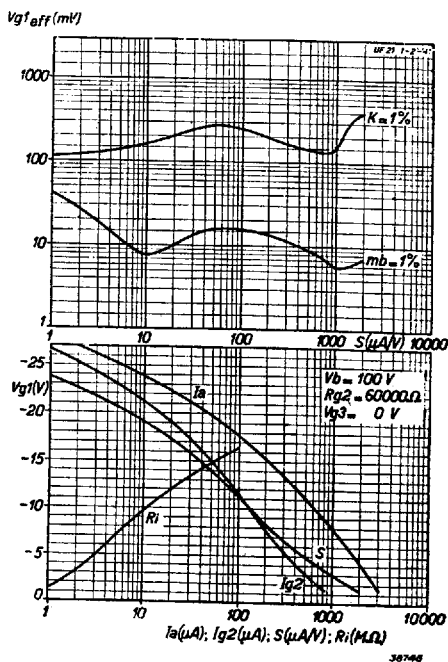


Fig. 8

At $V_b = 100 V$, $R_{g2} = 60,000$ Ohms (screen fed through a resistance) and $V_{g3} = 0 V$.

Upper diagram; maximum permissible effective value of alternating grid voltage with 1 % cross modulation ($K = 1\%$) and also with 1 % hum modulation ($mb = 1\%$), as function of mutual conductance.

Lower diagram; Mutual conductance, anode current, screen grid current and internal resistance as function of grid bias.

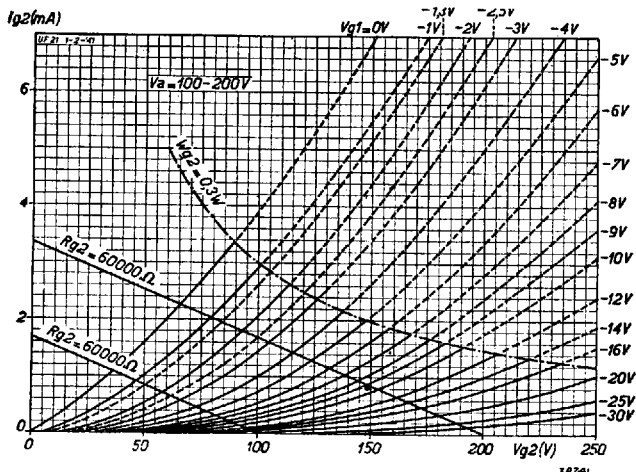


Fig. 9

Screen grid current as a function of screen voltage, with grid bias as parameter. The curves are a fair approximation in respect of all anode voltages between 100 and 200 V.